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Region-specific reliable channel estimation in RIS-enabled wireless communications via clustered federated learning^{*}

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ABSTRACT

Machine learning (ML)-based downlink channel estimation for reconfigurable intelligent surface (RIS)-assisted communication faces challenges such as handling channel variations, high communication overhead of centralized learning (CL), and vulnerability to malicious users. We propose a novel approach integrating blockchain to enhance security by verifying registered users, autoencoder (AE)-based clustering to identify regions within the cell, and clustered federated learning (CFL) to ensure good channel estimation performance while minimizing communication and energy overhead. Simulations show that the proposed clustering-based scheme achieves estimation performance comparable to CL while significantly reducing communication and energy overhead.

1. Introduction

Reconfigurable intelligent surfaces (RIS) are among the key enablers of the advancements shaping the next-generation wireless systems [1-3]. RIS consists of multiple small, low-cost, passive reflective elements, each intelligently controlled to impose independent phase shifts on incident signals [4-7]. When strategically deployed, RIS mitigates line-of-sight challenges, particularly in terahertz (THz) and millimeterwave (mmWave) communications [8,9]. However, realizing the full potential of RIS-enabled communication systems requires accurate downlink channel state information (CSI) [10], which poses a significant challenge due to the complexity of separately estimating channels from the base station (BS) to RIS and from RIS to the users. The complexity arises from the passive nature of RIS elements, requiring the estimation of the cascaded channel: while traditional techniques like least squares (LS) or minimum mean squared error (MMSE) can be used for cascaded channel estimation, the pilot overhead increases substantially in presence of RIS [11].

Machine learning (ML) techniques have effectively addressed channel estimation challenges: deep residual learning (DReL) was exploited to recover channel coefficients by learning residual noise [12]; CNNbased methods were considered to map low-dimensional cascaded channels to high-dimensional ones using a subset of RIS elements [13]; denoising NN-based cascaded channel estimation techniques were introduced in [14]. Although ML-based methods outperform traditional techniques, they rely on centralized frameworks requiring data transfer to a central server leading to high communication overhead and data integrity concerns [15]. Communication overhead can be reduced via federated learning (FL) where only model parameters (and not users data) are shared [16].

Despite the potential benefits, conventional FL-based approaches face challenges related to a single model not performing effectively across the whole cell due to variations in channel characteristics as users move between regions. Cell partitioning based on prior knowledge was explored in [11], but apparently limited by prior identification of the regions and does not include security mechanisms against malicious users. In our previous work, clustering was exploited for effective user selection, but the approach remained focused on building a single model [17].

This paper proposes a framework for downlink channel estimation in RIS-enabled communication systems based on clustered federated learning (CFL). More specifically, the proposed algorithm includes: (i) an autoencoder (AE)-based unsupervised region identification algorithm to group users based on the underlying characteristics of their received pilot signals, eliminating the need for prior knowledge; (ii) a simplified model-parameter selection which does not require an explicit classifier training to identify user transitions between regions; and (iii) a blockchain-based authentication mechanism that enhances security

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Fig. 1. Wireless communication through RIS assistance.

by ensuring only trustworthy users already connected to the BS participate in CFL training. Simulation results demonstrate the effectiveness of the unsupervised CFL in reducing communication and computational overhead without compromising system accuracy. Additionally, we investigate the impact of single-user poisoning within a cluster.

Notation. Uppercase (resp. lowercase) bold letters denote matrices (resp. column vectors). $(\cdot)^T$, \otimes , and $\|\cdot\|_2$ denote the transpose operator, the Kronecker product, and the Euclidean norm, respectively. diag(**a**) denotes a diagonal matrix with **a** on the main diagonal. The gradient of a function is denoted by $\nabla f(\cdot)$. Calligraphic letters denote sets (except \mathcal{L} denoting the loss function). |C| is the cardinality of C.

2. System model

2.1. Channel model

We consider a single-cell, narrowband, RIS-aided, wireless communication system, where the BS communicates with *K* single-antenna users using an *M*-antenna uniform planar array (UPA). A RIS with *N* reflecting elements is positioned between the BS and the users to improve communications and the BS controls the RIS through a dedicated wireless link [9,18,19]. The cell is segmented into *R* regions, see Fig. 1, each accommodating a specific number of users (U_r such that $\sum_{r=1}^{R} U_r = K$). Each region exhibits different specific channel properties tailored to its users and represents a different channel scenario influenced by factors such as the distance between the users and the BS, the angles of arrival/departure of the channels, etc. The transition from one region to another may cause significant changes in the channel characteristics experienced by the user [11].

We consider the frequency division duplex (FDD) mode for the RIS-aided system, similar to [20,21], where the downlink and uplink channels exhibit non-reciprocity.¹ Note that we focus solely on obtaining the CSI for the reflecting link. Meanwhile, the direct channels between the BS and the users can be estimated by turning off all RIS elements,² similar to conventional massive MIMO systems [11].

Neglecting the effect of the direct channel and assuming that the BS communicates with the users via the RIS reflecting link, the downlink signal $(y_{r,u})$ received at the *u*th user in the *r*th region is

$$y_{r,u} = \mathbf{f}_{r,u}^T \Phi \mathbf{G} \mathbf{w}_{r,u} x_{r,u} + n_{r,u}, r = 1, \dots, R, u = 1, \dots, U_r,$$
(1)

where $\mathbf{G} \in \mathbb{C}^{N \times M}$ represents the channel between the BS and the RIS, $\mathbf{f}_{r,u} \in \mathbb{C}^{N \times 1}$ denotes the channel between the RIS and the *u*th user in the *r*th region, and $n_{r,u} \in \mathbb{C}$ represents the additive white Gaussian noise. Further, $\mathbf{w}_{r,u} \in \mathbb{C}^{M \times 1}$ and $x_{r,u} \in \mathbb{C}$ denote the precoding vector at the BS and the transmitted signal by the BS, respectively. The reflecting matrix at the RIS, denoted by $\Phi \in \mathbb{C}^{N \times N}$, is diagonal, with $\Phi = \text{diag}(\phi_1, \phi_2, \dots, \phi_N)$, where ϕ_n is the reflection coefficient of the *n*th element of the RIS. Using the Saleh-Valenzuela model, the channel **G** can be expressed as

$$\mathbf{G} = \sqrt{\frac{MN}{L_G}} \sum_{l_1=1}^{L_G} \alpha_{l_1}^G \mathbf{a} \left(\vartheta_{l_1}^{G_r}, \psi_{l_1}^{G_r} \right) \mathbf{b} \left(\vartheta_{l_1}^{G_t}, \psi_{l_1}^{G_t} \right)^T,$$
(2)

where $\alpha_{l_1}^G$ signifies the complex gain, $\vartheta_{l_1}^{G_r}(\psi_{l_1}^{G_r})$ and $\vartheta_{l_1}^{G_t}(\psi_{l_1}^{G_t})$ denote the azimuth (elevation) angles at the RIS and the BS for the l_1 th path, respectively, and L_G is the number of paths between the RIS and the BS. Further, **G** remains nearly constant once the RIS is installed. Likewise, the channel $\mathbf{f}_{r,u} \in \mathbb{C}^{N \times 1}$ can be given as

$$\mathbf{f}_{r,u} = \sqrt{\frac{N}{L_{r,u}}} \sum_{l_2=1}^{L_{r,u}} \alpha_{l_2}^{r,u} \mathbf{a} \left(\vartheta_{l_2}^{r,u}, \psi_{l_2}^{r,u} \right), \tag{3}$$

where $L_{r,u}$ is the number of paths between the *u*th user in the *r*th region and the RIS. The variables $a_{l_2}^{r,u}$, $\vartheta_{l_2}^{r,u}$, and $\psi_{l_2}^{r,u}$ denote the complex gain, the azimuth angle, and the elevation angle associated with the l_2 path of the RIS, respectively. Further, $\mathbf{b}(\vartheta, \psi) \in \mathbb{C}^{M \times 1}$ and $\mathbf{a}(\vartheta, \psi) \in \mathbb{C}^{N \times 1}$ correspond to the normalized steering vectors linked with the BS and the RIS, respectively. For a standard $N_1 \times N_2$ UPA, the expression for $\mathbf{a}(\vartheta, \psi)$ can be given as [26]

$$\mathbf{a}(\vartheta,\psi) = \frac{1}{\sqrt{N}} \left[e^{-j\eta\cos(\psi)\mathbf{n}_1} \right] \otimes \left[e^{-j\eta\sin(\psi)\cos(\vartheta)\mathbf{n}_2} \right],\tag{4}$$

where $\eta = 2\pi d/\lambda$, $N = N_1 N_2$, $\mathbf{n}_1 = [0, 1, \dots, N_1 - 1]^T$ and $\mathbf{n}_2 = [0, 1, \dots, N_2 - 1]^T$, λ denotes the carrier wavelength and the antenna spacing *d* is typically chosen as $d = \lambda/2$. Defining $\boldsymbol{\phi} = [\phi_1, \phi_2, \dots, \phi_N]^T \in \mathbb{C}^{N \times 1}$, the system model in (1) can be remodeled as

$$y_{r,u} = \boldsymbol{\phi}^T \operatorname{diag}(\mathbf{f}_{r,u}^T) \mathbf{G} \mathbf{w}_{r,u} x_{r,u} + n_{r,u},$$
$$= \boldsymbol{\phi}^T \mathbf{H}_{r,u} \mathbf{w}_{r,u} x_{r,u} + n_{r,u},$$
(5)

¹ Even in the time division duplex (TDD) mode, the uplink and downlink channels may not be reciprocal in certain hardware implementations of RIS, as explicitly stated in [22]. Accurate downlink channel estimation is required to achieve effective beamforming for both the BS and the RIS [23].

² A practical method for turning off the RIS is proposed in [24,25].



Fig. 2. Proposed framework for cascaded downlink channel estimation.

where the downlink cascaded channel $\mathbf{H}_{r,u} \in \mathbb{C}^{N \times M}$ for the *u*th user in the *r*th region is defined as $\mathbf{H}_{r,u} = \text{diag}(\mathbf{f}_{r,u}^T)\mathbf{G}$. Due to the inability of the RIS to process the signals independently, the cascaded channel $\mathbf{H}_{r,u}$ is typically estimated rather than distinct channels \mathbf{G} and $\mathbf{f}_{r,u}$.

For the downlink cascaded channel estimation, the BS transmits known pilot signals to the users through the RIS over Q time slots. Using (5), the pilot signal received at the *u*th user in the *r*th region during the *q*th time slot, denoted by $y_{r,u,q}^{\rho} \in \mathbb{C}$, can be expressed as

$$y_{r,u,q}^{p} = \boldsymbol{\phi}_{q}^{T} \mathbf{H}_{r,u} \mathbf{w}_{r,u} p_{r,u,q} + n_{r,u,q},$$
(6)

where q = 1, 2, ..., Q. Further, $p_{r,u,q}$, ϕ_q and $n_{r,u,q} \sim C\mathcal{N}(0, \sigma_n^2)$ denote the pilot signal transmitted by the BS, the reflection vector at the RIS and the noise at the user associated with the *q*th time slot, respectively. Considering *Q* time slots, the received pilot vector $\mathbf{y}_{r,u}^p$ at the *u*th user can be given as

$$\mathbf{y}_{r,u}^{p} = \boldsymbol{\Theta} \mathbf{H}_{r,u} \mathbf{w}_{r,u} + \mathbf{n}_{r,u},\tag{7}$$

where $\mathbf{y}_{r,u}^p = [y_{r,u,1}^p, y_{r,u,2}^p, \dots, y_{r,u,Q}^p]^T \in \mathbb{C}^{Q \times 1}$, the pilot signal is chosen as $p_{r,u,q} = 1, \boldsymbol{\Theta}^T = [\boldsymbol{\phi}_1, \boldsymbol{\phi}_2, \dots, \boldsymbol{\phi}_Q] \in \mathbb{C}^{N \times Q}$ and $\mathbf{n}_{r,u} = [n_{r,u,1}, n_{r,u,2}, \dots, n_{r,u,Q}]^T \in \mathbb{C}^{Q \times 1}$. Using the identity vec(ABC) = ($\mathbf{C}^T \otimes \mathbf{A}$)vec(B), the above

system model can be modified as

$$\mathbf{y}_{r,u}^{p} = (\mathbf{w}_{r,u}^{T} \otimes \boldsymbol{\Theta}) \operatorname{vec}(\mathbf{H}_{r,u}) + \mathbf{n}_{r,u},$$
$$= \boldsymbol{\Psi}_{r,u} \mathbf{h}_{r,u} + \mathbf{n}_{r,u},$$
(8)

where $\Psi_{r,u} = (\mathbf{w}_{r,u}^T \otimes \Theta) \in \mathbb{C}^{Q \times MN}$ and $\mathbf{h}_{r,u} = \operatorname{vec}(\mathbf{H}_{r,u}) \in \mathbb{C}^{MN \times 1}$. It is assumed that Θ and $\mathbf{w}_{r,u}$ remain fixed and are known to both the BS and the user. The primary goal is to estimate $\mathbf{h}_{r,u}$ using $\mathbf{y}_{r,u}^p$ and $\Psi_{r,u}$.

3. Proposed algorithm

The proposed framework consists of two main blocks (see Fig. 2). The upper block includes authentication, training, classification, and CFL-based channel estimation models. The lower block leverages these models to categorize input pilot signals into specific regions, and subsequently employs the designated region models for channel estimation, akin to the testing phase.

3.1. Blockchain-based authentication

Authentication is implemented to establish a secure environment and ensure the participation of trustworthy users only. In traditional FL-based schemes for CSI estimation, the users are typically authenticated to confirm their registration and connectivity with the BS. However, this process could allow adversaries to connect to the BS and poison the model parameters merely by participating in the network. To address this vulnerability, a blockchain-based authentication mechanism has been implemented on top of the traditional system for CFL training, thus trustworthy users participate in the training, while other users receive the final trained model for channel estimation purposes. Blockchain technology is selected due to four crucial benefits: (i) data integrity through cryptographic features, (ii) user privacy, (iii) distributed structure robust to single point of failure, and (iv) public records for fair and transparent auditing [27-29].

A smart contract on the Ethereum platform handles user authentication and registration performing two main functions: (i) registering trustworthy users based on the hash of their identity, and (ii) authenticating users before they participate in CFL model training. Each trustworthy user is assigned a unique identifier for registration, such as an international mobile subscriber identifier (IMSI), coupled with a secret phrase, both managed by a designated authority.³ The authority is the only entity authorized to update the list of registered users. When a user connects to the BS, the BS hashes the IMSI along with the secret phrase to confirm the user's registration. Since this is a onetime process, sharing the secret imposes minimal to no overhead. Once registration is confirmed, the user can participate in the CFL-based training. Furthermore, the blockchain is managed by the BS, which has sufficient computational resources to handle them efficiently.

3.2. Unsupervised clustering and classification

Each authenticated user trains a lightweight AE using its received signals for a limited number of epochs. It is crucial to note that at this stage, the BS and the users operate without prior knowledge of the regions. Each user subsequently shares its encoder/decoder model parameters and 5% of its received pilot signals with the BS. The BS tests data from each user against all AE parameters and generates a loss matrix where each row serves as user embedding, capturing the loss values corresponding to user's data when processed by all other users' AEs. The embeddings are then clustered using K-means algorithm, where the optimal number of clusters is determined using the elbow method.⁴ This regional information is then utilized to train tailored region-specific models.

The BS selects one user per cluster⁵ and assumes the AE parameters of the selected user as representative for the corresponding region. The set of selected AE parameters (one per region) allows for effective assignment of channel-estimation models to mobile users, who can potentially transit from one region into a different region, and to new users joining the system. The effective assignment relies on classification based on the minimum reconstruction loss of the reference AE parameters operating on signals from the user of interest. It is worth noticing that the classifier is built on the classes identified by the clustering algorithm and thus requires no prior information. The overall methodology of this clustering process is detailed in Algorithm 1.

Algorithm 1 AE-based Clustering and Classification

1: procedure TRAIN-AE for *k*th user, $k = 1, \ldots, K$, do 2: 3: Initialize β, γ $\min_{\boldsymbol{\gamma},\boldsymbol{\beta}} \left(\frac{1}{s} \sum_{i=1}^{s} \| \mathbf{x}_{k,i} - \hat{\mathbf{x}}_{k,i} \|_{2}^{2} \right) \\ AE_{k} = [\boldsymbol{\beta}, \boldsymbol{\gamma}] \\ \text{Share } AE_{k} \text{ and } D_{k}^{\text{AE}} \text{ with the BS}$ 4: 5: 6: 7: end for 8: end procedure 9: procedure Loss MATRIX (LM) Initialize LM with zeros 10.

11: for each AE_i , $i = 1, \dots, K$, do

for each D_i^{AE} , $j = 1, \dots, K$, do 12:

$$LM_{i,j} = \mathcal{L}(\mathbf{X}, \hat{\mathbf{X}}), \text{ where } \begin{cases} \mathbf{X} \in D_j^{AE}, \\ \hat{\mathbf{X}} = AE_i(\mathbf{X}) \end{cases}$$

14:

13:

15: end for

16: end for

17: end procedure

- 18: procedure Clustering
- 19: $\mathcal{Z} = \mathrm{EM}(LM)$

 \triangleright *EM* represents the Elbow Method. $Clts = KM(LM, \mathcal{Z})$ \triangleright *Clts* = {*C*₁, *C*₂, ..., *C*_Z} is the set of 20: clusters, and KM represents the K-means algorithm.

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21: end procedure
```

22: procedure CLASSIFICATION

23: $AE_{\text{Classify}} = \{f(C_1), f(C_2), ..., f(C_z)\} \triangleright AE_{\text{Classify}} \text{ represents the}$ AE-based classification of the received signals, and $f(C_i)$ selects an AE from cluster C_i .

24: Loss = []

for each AE in AE_{Classify} do 25:

26: $\mathcal{L}_{AE} = \mathcal{L}(\mathbf{y}^p, AE(\mathbf{y}^p))$

 $Loss \leftarrow Loss \cup \{\mathcal{L}_{AF}\}$ 27:

28: end for

 $I = Index(\min(Loss))$ 29:

30: end procedure

3.3. Communication and energy efficient training

After identifying different regions within a cell, the CFL optimizes the parameters for each identified region/cluster using the collective contributions of the corresponding users. During each communication round, the BS selects a subset (S) of users from a region and provides region-specific model parameters to solve the following optimization problem [33]:

$$\min_{\theta_r} \underbrace{\frac{1}{|S|} \sum_{u \in S} \mathcal{L}_k(\theta_r; D_u)}_{\mathcal{L}(\theta; D)}, \tag{9}$$

where the vector θ_r collects the globally shared model parameters within a region and $\mathcal{L}_k(\theta; \mathcal{D}_u)$ is the local objective function used in conjunction with the local dataset D_u for the *u*th client. The users refine the parameters using their local datasets over multiple epochs with a batch size of B and send the updated parameters back to the BS. The BS then aggregates the refined models and updates the associated cluster models. This iterative process continues until convergence or a predetermined number of communication rounds is reached.

Unlike traditional approaches [11,16], the proposed framework involves only a subset of users per communication round, significantly reducing communication load and the BS's computational burden for parameter aggregation. Random user selection further conserves energy by preventing continuous user activity. The CFL training algorithm is summarized in Algorithm 2.

³ It refers to the administrative entity within the mobile subscriber company or the organization responsible for the framework's design. Furthermore, user trust for registration can be determined by analyzing network traffic and user behavior [30-32].

⁴ Ideally, the number of clusters equals the number of regions (*R*).

⁵ The impact of the user selection strategy (e.g. random selection, optimized strategy) falls beyond the scope of this work.

Algorithm 2 Clustered Federated Learning

1:	procedure Federated Learning
2:	Initialize: Global model weights $\theta_{r,0}$
3:	for each round $n = 1, 2,$ do
4:	$S_n \leftarrow \text{RandomSelect}(U_r, m) \triangleright \text{Selecting } m \text{ users from } U_r$
	users corresponding to the <i>r</i> th region
5:	for each <i>u</i> th user in S_n do
6:	$\theta_{r,u,n} \leftarrow \text{ClientUpdate}(\mathcal{D}_{r,u}, \theta_{r,n})$
7:	end for
8:	$\theta_{r,n+1} \leftarrow \frac{1}{ S_n } \sum_{u \in S_n} \theta_{r,u,n}$
9:	end for
10:	end procedure
11:	procedure ClientUpdate($D_{r,u}, \theta_r$)
12:	$\mathcal{B}_k \leftarrow Batches(\mathcal{D}_{r,u}, B)$
13:	$\theta_{r,u} \leftarrow \theta_r$
14:	for $e = 1$ to \mathcal{E} do $\triangleright \mathcal{E}$ represents the number of epochs
15:	for each $b \in \mathcal{B}_k$ do
16:	$\boldsymbol{\theta}_{r,u} \leftarrow \boldsymbol{\theta}_{r,u} - \eta \nabla \mathcal{L}_{r,u}(\boldsymbol{\theta}_{r,u}; b)$
17:	end for
18:	end for
19:	Share $\theta_{r,u}$ with the BS
20:	end procedure

3.4. Communication overhead

Communication overhead (O^C) refers to data exchange during the training phase of the model. The communication overhead in centralized learning (CL) is characterized by the number of data symbols transmitted by the user to the BS for training the centralized model, comprising both input samples and their corresponding outputs. Conversely, the communication overhead for CFL, is defined as the total number of parameters transmitted by the user and received from the BS during the CFL training [16].

3.5. Energy overhead

Energy overhead (O^E) refers to the number of iterations required to update the model parameters during training. The energy overhead in CL is the product of the number of batches and the number of epochs. Conversely, the energy overhead in FL is characterized by the product of the number of batches per user, the number of epochs per user, the number of selected users per communication round, and the total number of communication rounds.⁶

4. Simulation results

This section presents the simulation results to validate the performance of our proposed algorithms. We consider a single-cell, RIS-assisted communication system, where a BS with M = 16 antenna serves K = 15 single-antenna users using a N = 64 element UPA RIS. The cell is segmented into R = 3 regions, where each region has different channel characteristics and contains $U_r = 5$ users for r = 1, 2, 3. The channel matrix G between the BS and the RIS is generated using Eq. (2) with $L_G = 3$ paths between the BS and the RIS. The complex gains are modeled as complex Gaussian random variables with zero mean and unit variance, *i.e.* $\alpha_{l_1}^G \sim C\mathcal{N}(0, 1)$, and the phase angles are uniformly distributed over $(-\pi/2, \pi/2)$. Each element of $\Psi_{r,u}$ in Eq. (8) is chosen from the set $\left\{-\frac{1}{\sqrt{Q}}, \frac{1}{\sqrt{Q}}\right\}$, considering discrete phase shifts at

Table 1		
Autoencoder	layer	architecture.

Layer type Input size		Output size	
Lincor	256	E10	
Lineai	230	312	
Linear	512	256	
Linear	256	128	
Linear	128	64	
Linear	64	32	
Linear	32	16	
Linear	16	32	
Linear	32	64	
Linear	64	128	
Linear	128	256	
Linear	256	512	
Linear	512	256	

Table	2
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Personalized cluster model architecture.	
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Layer type	Filter size	Output channels
Convolutional Layer 1	3 × 3	32
Batch Normalization Layer	-	-
Convolutional Layer 2	3×3	16
Batch Normalization Layer	-	-
Convolutional Layer 3	3×3	8
Batch Normalization Layer	-	-
Fully Connected Layer	-	Output

the BS and the RIS, and $\Psi_{r,u}$ is assumed to be identical $\forall u, r$. The signalto-noise ratio (SNR) is defined as $1/\sigma_n^2$, where σ_n^2 is the noise variance. We employ the Saleh-Valenzuela channel model for the channel $\mathbf{f}_{r,u}$ in Eq. (3) from RIS to the *u*th user in the *r*th region. The path count between the RIS and the user is $L_{r,u} = 3$, $\forall u, r$. The complex channel gains $a_{l_2}^{r,u}$ follow the complex Gaussian distribution, *i.e.*, $a_{l_2}^{r,u} \sim C\mathcal{N}(0,1)$ and phase angles are uniformly distributed over $(-\pi/2, \pi/2)$. The elevation angles are equally divided into three segments $(-\pi/2, -\pi/6)$, $(-\pi/6, \pi/6)$, and $(\pi/6, \pi/2)$, with each segment corresponding to a specific region and channel characteristics. The operating frequency is 28 GHz.

The required pilot overhead is set as $Q = \frac{NM}{8} = 128$ [11]. Each user collects a dataset of $|D_{r,u}| = 20,000$ samples, with 10% reserved for testing. Each user is equipped with an AE for clustering and classification. The received pilot signal is transformed into a real-valued vector by concatenating its real and imaginary parts, with the first half containing the real part and the second half containing the imaginary part. The AE's takes this as input and maps it to 512 features, then progressively reduces the feature size by half in each step until reaching 16. It then reconstructs the data by symmetrically increasing the feature size back to the original input dimensions. Table 1 outlines the AE architecture. The AE is trained using the Adam optimizer, with a learning rate of $\eta = 1 \times 10^{-3}$ and a batch size of B = 1024 in PyTorch.

Each personalized cluster model employs a CNN architecture with three 3×3 convolutional layers, producing 32, 16, and 8 channels, followed by batch normalization and ReLU activation. The output is then flattened and passed through a fully connected layer, transforming into a 2048-dimensional feature vector. The complete model includes approximately 2.1 million trainable parameters, as detailed in Table 2. We analyzed five scenarios to evaluate the impact of selecting the optimal number of users and local training epochs per communication round in CFL: in scenario S1, all five users in a cluster participate, each running one local training epoch; in scenario S2, three out of five users per round are selected, each training for one epoch; in scenario S3 two users per round are randomly selected, each training for one epoch; in scenario S4, two randomly chosen users train for three epochs each; and in scenario S5, a single user is selected to train for five epochs. The ML model architecture remains unchanged across all configurations.

Fig. 3 presents the loss metrics by evaluating each user's data against AE models trained on individual users. The vertical axis represents tested users, while the horizontal axis corresponds to AE models.

⁶ If all users participate in each round, FL experiences the same energy overhead as CL.







Fig. 4. Clustering and classification performance.



Fig. 5. Performance comparison.

The AE's ability to regenerate similar signal structures is utilized for clustering, where similar loss embeddings reflect shared user characteristics. As shown in Fig. 3, an AE model trained on one user's data typically yields lower reconstruction loss on data that shares similar statistical properties. In contrast, when the AE is applied to data from users in different clusters (*e.g.*, with differing environmental conditions), the reconstruction loss tends to be higher. Finally, the *K*-means algorithm, applied to the loss metrics, identifies user clusters, enabling the training of a cluster-specific model for each cluster.

The within-cluster sum of squares (WCSS) is calculated to determine the optimal number of clusters, as shown in Fig. 4(a). Lower WCSS values indicate tighter clusters with data points closer to their centroids. For the classifier, we consider random selection in each cluster. Fig. 4(b) shows the classifier's efficiency on 30,000 test samples, maintaining 97 \pm 1% accuracy without additional training. This showcases the classifier's ability to correctly identify cluster and select the appropriate cluster-specific model for channel estimation.

Fig. 5(a) compares CFL with two traditional FL approaches both training a single global model: one based on randomly selecting three users from each region, and the other using all available users. Conversely, CFL trains region-specific models based on user clustering. Traditional LS and MMSE methods are also shown as benchmarks; the required pilot overhead for LS and MMSE is set at Q = NM = 1024. Apparently, the proposed approach significantly outperforms the traditional benchmarks and the conventional FL alternatives, demonstrating its superior performance.

Fig. 5(b) analyzes the performance of various strategies on a test dataset of 30,000 samples that cover the entire cell, ensuring equal contributions from each cluster. Testing involves classifying input samples into their regions before applying specific channel estimation models.



Fig. 6. Comparison of communication and energy overhead.

Table 3

Secured vs. unsecured at $SNR = 15$ dB.					
	S1	S2	S3	S4	S5
Secured (dB)	-15.55	-15.31	-15.28	-15.43	-15.40
Unsecured (dB)	-2.69	-1.43	-10.57	-12.86	-12.47

Results demonstrate that selecting a smaller subset of users within the cluster and fewer communication rounds maintains performance instead of utilizing all users in traditional FL and centralized methods for channel estimation. The proposed CFl based strategy even slightly enhances the performance, particularly in scenario S4 at 15 dB, compared to the all-users scenario.

Fig. 6(a) presents a comparative analysis of various training configurations, focusing on communication overhead of CL, conventional FL, and proposed CFL approaches. It is apparent that CL is more efficient than conventional FL in all scenarios and surpasses the proposed CFL in scenarios S1, S2, and S3, while the CFL⁷ is preferable in scenarios S4 and S5. Fig. 6(b) the energy overhead during training for channel estimation, showing that effective user selection significantly impacts the overall energy consumption. Additionally, energy utilization per user decreases as the probability of selecting a specific user drops from 1 to 0.2 when considering a single user. Overall, Fig. 6 validate that the proposed CFL-based strategy achieves strong performance while significantly reducing communication and computational overhead, highlighting its effectiveness and efficiency for scalable deployment.

Table 3 compares two CFL systems: one with authentication, ensuring all five users are non-malicious, and another allowing all users, including one malicious user. The malicious user, once selected, receives the model from the BS, ignores the learning process, randomly reinitializes the parameters, and returns them to the BS. These actions disrupt training and worsen NMSE, especially in scenarios S1 and S2, where the adversarial user is frequently selected. Differently, in scenarios S4 and S5 the chances of excluding the adversarial user increase and system performance improves. Generally, the system performs optimally when all users are authenticated and trustworthy, ensuring accurate channel estimation.

5. Conclusion and future work

This paper proposed a CFL-based framework enhancing communication efficiency, energy efficiency, and security. We designed an unsupervised clustering and classification method based on the received pilot signals, which enhances the channel estimation accuracy since users within the same cluster can collaborate more efficiently. We demonstrated that strategically selecting fewer users and optimizing the number of local epochs can significantly improve communication efficiency and energy efficiency. Additionally, we employed blockchain technology to add a layer of protection and prevent malicious users from compromising the model. For future work, we aim to explore scenarios involving user mobility and various adversarial situations.

CRediT authorship contribution statement

Muhammad Asaad Cheema: Writing – review & editing, Writing – original draft, Software, Methodology, Formal analysis, Conceptualization. **Apoorva Chawla:** Writing – review & editing, Writing – original draft, Conceptualization. **Vinay Chakravarthi Gogineni:** Writing – review & editing, Conceptualization. **Pierluigi Salvo Rossi:** Writing – review & editing, Supervision, Funding acquisition, Conceptualization.

Declaration of competing interest

The authors declare the following financial interests/personal relationships which may be considered as potential competing interests: Muhammad Asaad Cheema reports financial support was provided by Research Council of Norway. If there are other authors, they declare that they have no known competing financial interests or personal relationships that could have appeared to influence the work reported in this paper.

Data availability

No data was used for the research described in the article.

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 $^{^7\,}$ Note that our approach accounts for the overhead of sharing 5% of data and AE parameters for clustering and classification.

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